

50 A VRPower[®] Integrated Power Stage

DESCRIPTION

The SiC788 and SiC788A are integrated power stage solutions optimized for synchronous buck applications to offer high current, high efficiency, and high power density performance. Packaged in Vishay's proprietary 6 mm x 6 mm MLP package, SiC788 and SiC788A enable voltage regulator designs to deliver up to 50 A continuous current per phase.

The internal power MOSFETs utilize Vishay's state-of-the-art Gen IV TrenchFET technology that delivers industry benchmark performance to significantly reduce switching and conduction losses.

The SiC788 and SiC788A incorporate an advanced MOSFET gate driver IC that features high current driving capability, adaptive dead-time control, an integrated bootstrap Schottky diode, a thermal warning (THWn) that alerts the system of excessive junction temperature, and skip mode (SMOD#) to improve light load efficiency. The drivers are also compatible with a wide range of PWM controllers and supports tri-state PWM, 3.3 V (SiC788A) / 5 V (SiC788) PWM logic.

FEATURES

diode

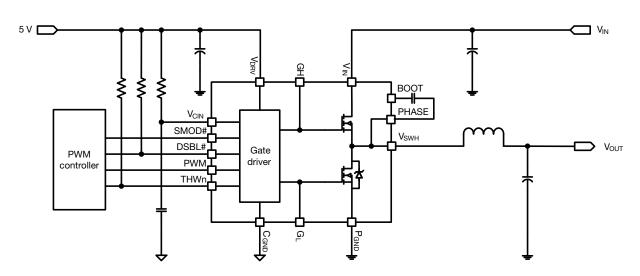
- Thermally enhanced PowerPAK[®] MLP66-40L package
- Package
 Vishay's Gen IV MOSFET technology and a collow-side MOSFET with integrated Schottky



- Delivers up to 50 A continuous current
- 95 % peak efficiency
- High frequency operation up to 1.5 MHz
- Power MOSFETs optimized for 12 V input stage
- 3.3 V (SiC788A) / 5 V (SiC788) PWM logic with tri-state and hold-off
- SMOD# logic for light load efficiency improvement
- Low PWM propagation delay (< 20 ns)
- Thermal monitor flag
- Faster enable / disable
- Under voltage lockout for V_{CIN}
- Material categorization: for definitions of compliance please see <u>www.vishay.com/doc?99912</u>

APPLICATIONS

• Multi-phase VRDs for CPU, GPU, and memory



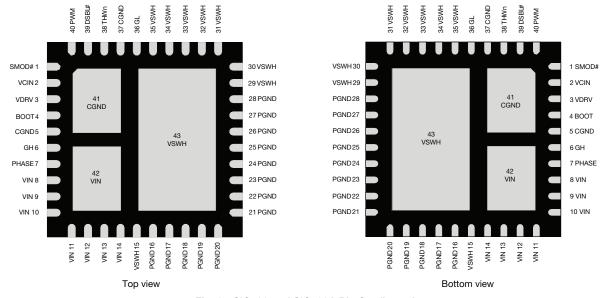
TYPICAL APPLICATION DIAGRAM

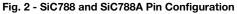
Fig. 1 - SiC788 and SiC788A Typical Application Diagram

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PINOUT CONFIGURATION





PIN DESCRIPTIO	IN DESCRIPTION				
PIN NUMBER	NAME	FUNCTION			
1	SMOD#	Low-side gate turn-off logic. Active low			
2	V _{CIN}	Supply voltage for internal logic circuitry			
3	V _{DRV}	Supply voltage for internal gate driver			
4	BOOT	High-side driver bootstrap voltage			
5, 37, 41	C _{GND}	Analog ground for the driver IC			
6	GH	High-side gate signal			
7	PHASE	Return path of high-side gate driver			
8 to 14, 42	V _{IN}	Power stage input voltage. Drain of high-side MOSFET			
15, 29 to 35, 43	V _{SWH}	Switch node of the power stage			
16 to 28	P _{GND}	Power ground			
36	GL	Low-side gate signal			
38	THWn	Thermal warning open drain output			
39	DSBL#	Disable pin. Active low			
40	PWM	PWM control input			

ORDERING INFORMATION						
PART NUMBER	PACKAGE	MARKING CODE	OPTION			
SiC788ACD-T1-GE3	PowerPAK [®] MI P66-40I	SiC788A	3.3 V PWM optimized			
SiC788CD-T1-GE3	FOWEIFAK* MILF00-40L	SiC788	5 V PWM optimized			
SiC788ADB and SiC788DB		Reference board				



ABSOLUTE MAXIMUM RATING	GS			
ELECTRICAL PARAMETER	CONDITIONS	LIMIT	UNIT	
Input Voltage	V _{IN}	-0.3 to +25		
Control Logic Supply Voltage	V _{CIN}	-0.3 to +7		
Drive Supply Voltage	V _{DRV}	-0.3 to +7		
Switch Node (DC voltage)	N N	-0.3 to +25		
Switch Node (AC voltage) (1)	V _{SWH}	-8 to +30	1	
BOOT Voltage (DC voltage)		32	V	
BOOT Voltage (AC voltage) (2)	V _{BOOT}	38		
BOOT to PHASE (DC voltage)		-0.3 to +7		
BOOT to PHASE (AC voltage) (3)	VBOOT- PHASE	-0.3 to +8	1	
All Logic Inputs and Outputs (PWM, DSBL#, and THWn)		-0.3 to V _{CIN} + 0.3		
Output Output I (4)	$f_{S} = 300 \text{ kHz}, V_{IN} = 12 \text{ V}, V_{OUT} = 1.8 \text{ V}$	50		
Output Current, I _{OUT(AV)} ⁽⁴⁾	f _S = 1 MHz, V _{IN} = 12 V, V _{OUT} = 1.8 V	40	A	
Max. Operating Junction Temperature	TJ	150		
Ambient Temperature	T _A	-40 to +125		
Storage Temperature	T _{stg}	-65 to +150	1	
Electrostatio Discharge Protection	Human body model, JESD22-A114	5000	V	
Electrostatic Discharge Protection	Charged device model, JESD22-C101	1000	v	

Notes

 $^{(1)}$ The specification values indicated "AC" is V_{SWH} to P_{GND}, -8 V (< 20 ns, 10 µJ), min. and 30 V (< 50 ns), max.

 $^{(2)}$ The specification value indicates "AC voltage" is V_{BOOT} to P_{GND}, 36 V (< 50 ns) max.

⁽³⁾ The specification value indicates "AC voltage" is V_{BOOT} to V_{PHASE}, 8 V (< 20 ns) max.

(4) Output current rated with testing evaluation board at T_A = 25 °C with natural convection cooling. The rating is limited by the peak evaluation board temperature, T_J = 150 °C, and varies depending on the operating conditions and PCB layout. This rating may be changed with different application settings.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING RANGE							
ELECTRICAL PARAMETER	MINIMUM TYPICAL		MAXIMUM	UNIT			
Input Voltage (V _{IN})	4.5	-	18				
Drive Supply Voltage (V _{DRV})	4.5	5	5.5	V			
Control Logic Supply Voltage (V _{CIN})	4.5	5	5.5	v			
BOOT to PHASE (V _{BOOT-PHASE} , DC voltage)	4	4.5	5.5				
Thermal Resistance from Junction to PAD	-	1	-	°C/W			
Thermal Resistance from Junction to Case	-	2.5	-	C/W			

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			LIMITS				
PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.		
POWER SUPPLY			•	•	•		
		V _{DSBL#} = 0 V, no switching, V _{PWM} = FLOAT	-	85	-		
Control Logic Supply Current	I _{VCIN}	V _{DSBL#} = 5 V, no switching, V _{PWM} = FLOAT	-	290	-	μA	
		V _{DSBL#} = 5 V, f _S = 300 kHz, D = 0.1	-	295	-		
		f _S = 300 kHz, D = 0.1	-	9	15	mA	
Drive Supply Current	h	f _S = 1 MHz, D = 0.1	-	30	-		
Silve Supply Sulten	I _{VDRV}	$V_{DSBL\#} = 0 V$, no switching	-	30	-	μA	
		V _{DSBL#} = 5 V, no switching	-	55	-	μΛ	
BOOTSTRAP SUPPLY			1	1		L	
Bootstrap Diode Forward Voltage	V _F	$I_F = 2 \text{ mA}$			0.4	V	
PWM CONTROL INPUT (SiC788)	1 · · · · · · · · · · · · · · · · · · ·		L		1	1	
Rising Threshold	V _{TH_PWM_R}		3.4	3.7	4.0		
alling Threshold	V _{TH_PWM_F}		0.72	0.9	1.1		
Tri-state Voltage	V _{TRI}	$V_{PWM} = FLOAT$	-	2.3	-	V	
ri-state Rising Threshold	V _{TRI_TH_R}		0.9	1.15	1.38	-	
ri-state Falling Threshold	V _{TRI_TH_F}		3.1	3.35	3.6	<u> </u>	
ri-state Rising Threshold lysteresis	V _{HYS_TRI_R}		-	225	-	mV	
ri-state Falling Threshold lysteresis	V _{HYS_TRI_F}		-	325	-		
PWM Input Current	I _{PWM}	V _{PWM} = 5 V	-	-	350	μA	
•		V _{PWM} = 0 V	-	-	-350	P.,	
WM CONTROL INPUT (SiC788A)				0.45		r —	
Rising Threshold	V _{TH_PWM_R}		2.2	2.45	2.7		
alling Threshold	V _{TH_PWM_F}		0.72	0.9	1.1		
ri-state Voltage	V _{TRI}	$V_{PWM} = FLOAT$	-	1.8	-	V	
ri-state Rising Threshold	V _{TRI_TH_R}		0.9	1.15	1.38		
ri-state Falling Threshold	V _{TRI_TH_F}		1.95	2.2	2.45		
ri-state Rising Threshold lysteresis	V _{HYS_TRI_R}		-	225	-	- m\	
ri-state Falling Threshold lysteresis	V _{HYS_TRI_F}		-	275	-		
PWM Input Current	I _{PWM}	V _{PWM} = 3.3 V	-	-	225	μA	
	PWM	$V_{PWM} = 0 V$	-	-	-225	μ/	
IMING SPECIFICATIONS			1	T	•	r —	
ri-State to GH/GL Rising ropagation Delay	t _{PD_TRI_R}		-	30	-		
ri-state Hold-off Time	t _{TSHO}		-	130	-		
iH - Turn Off Propagation Delay	t _{PD_OFF_GH}		-	18	-		
iH - Turn On Propagation Delay Dead time rising)	t _{PD_ON_GH}	No load, see fig. 4	-	15	-		
GL - Turn Off Propagation Delay	t _{PD_OFF_GL}		_	12	-	ns	
iL - Turn On Propagation Delay Dead time falling)	t _{PD_ON_GL}		-	8	-		
SBL# Low to GH/GL Falling ropagation Delay	t _{PD_DSBL#_F}	Fig. 5	-	15	-		
SBL# High to GH/GL Rising Propagation Delay	t _{PD_DSBL#_R}	Fig. 5	-	20	-		
WM Minimum On-time	t _{PWM_ON_MIN}		30	-	-		
SBL# SMOD# INPUT							
SPI # Logic Input Voltage	V _{IH_DSBL#}	Input logic high	2	-	-		
SBL# Logic Input Voltage	V _{IL_DSBL#}	Input logic low	-	-	0.8	v	
MOD# Logic Input Voltage	V _{IH_SMOD#}	Input logic high	2	-	-	v	
	VIL_SMOD#	Input logic low	-	-	0.8	1	

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ELECTRICAL SPECIFICAT (DSBL# = SMOD# = 5 V, V _{IN} =		and $V_{CIN} = 5 \text{ V}, \text{ T}_{A} = 25 \text{ °C})$					
PARAMETER	SYMBOL	TEST CONDITION		LIMITS		UNIT	
FARAMETER	SYMBOL TEST CONDITION		MIN.	TYP.	MAX.	UNIT	
PROTECTION							
	V _{UVLO}	V _{CIN} rising, on threshold	-	3.7	4.1	v	
Under Voltage Lockout		V _{CIN} falling, off threshold	2.7	3.1	-	v	
Under Voltage Lockout Hysteresis	V _{UVLO_HYST}		-	575	-	mV	
THWn Flag Set ⁽²⁾	T _{THWn_SET}		-	160	-		
THWn Flag Clear ⁽²⁾	T _{THWn_CLEAR}		-	135	-	°C	
THWn Flag Hysteresis ⁽²⁾	T _{THWn_HYST}		-	25	-		
THWn Output Low	V _{OL_THWn}	I _{THWn} = 2 mA	-	0.02	-	V	

Notes

⁽¹⁾ Typical limits are established by characterization and are not production tested.

⁽²⁾ Guaranteed by design.

DETAILED OPERATIONAL DESCRIPTION

PWM Input with Tri-state Function

The PWM input receives the PWM control signal from the VR controller IC. The PWM input is designed to be compatible with standard controllers using two state logic (H and L) and advanced controllers that incorporate tri-state logic (H, L and tri-state) on the PWM output. For two state logic, the PWM input operates as follows. When PWM is driven above V_{PWM TH B} the low-side is turned OFF and the high-side is turned ON. When PWM input is driven below VPWM TH F the high-side is turned OFF and the low-side is turned ON. For tri-state logic, the PWM input operates as previously stated for driving the MOSFETs when PWM is logic high and logic low. However, there is a third state that is entered as the PWM output of tri-state compatible controller enters its high impedance state during shut-down. The high impedance state of the controller's PWM output allows the SiC788 and SiC788A to pull the PWM input into the tri-state region (see definition of PWM logic and tri-state, fig. 4). If the PWM input stays in this region for the tri-state hold-off period, t_{TSHO}, both high-side and low-side MOSFETs are turned OFF. The function allows the VR phase to be disabled without negative output voltage swing caused by inductor ringing and saves a Schottky diode clamp. The PWM and tri-state regions are separated by hysteresis to prevent false triggering. The SiC788A incorporates PWM voltage thresholds that are compatible with 3.3 V logic and the SiC788 thresholds are compatible with 5 V logic.

Disable (DSBL#)

In the low state, the DSBL# pin shuts down the driver IC and disables both high-side and low-side MOSFETs. In this state, standby current is minimized. If DSBL# is left unconnected, an internal pull-down resistor will pull the pin to C_{GND} and shut down the IC.

Pre-Charger Function

When DSBL# is driven from below $V_{IL_DSBL\#}$ to above $V_{IH_DSBL\#}$ the low-side is turned ON for a short duration (60 ns typical) to refresh the BOOT capacitor in case it has been discharged due to the driver being in standby for a long period of time.

Diode Emulation Mode (SMOD#)

When SMOD# is logic low diode emulation mode is enabled and the low-side is turned OFF. This is a non-synchronous conversion mode that improves light load efficiency by reducing switching losses. Conducted losses that occur in synchronous buck regulators when inductor current is negative can also be reduced. Circuitry in the external controller IC detects when inductor current crosses zero and drives SMOD# below $V_{IL_SMOD#}$ turning the low-side MOSFET OFF. The function can be also be used for a pre-biased output voltage. If SMOD# is left unconnected, an internal pull up resistor will pull the pin to V_{CIN} (logic high) to disable the SMOD# function.

Thermal Shutdown Warning (THWn)

The THWn pin is an open drain signal that flags the presence of excessive junction temperature. Connect, with a maximum of 20 k Ω , to V_{CIN}. An internal temperature sensor detects the junction temperature. The temperature threshold is 160 °C. When this junction temperature is exceeded the THWn flag is set. When the junction temperature drops below 135 °C the device will clear the THWn signal. The SiC788 and SiC788A do not stop operation when the flag is set. The decision to shutdown must be made by an external thermal control function.

Voltage Input (VIN)

This is the power input to the drain of the high-side power MOSFET. This pin is connected to the high power intermediate BUS rail.

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Switch Node (V_{SWH} and PHASE)

The switch node, V_{SWH} , is the circuit power stage output. This is the output applied to the power inductor and output filter to deliver the output for the buck converter. The PHASE pin is internally connected to the switch node, V_{SWH} . This pin is to be used exclusively as the return pin for the BOOT capacitor. A 20 k Ω resistor is connected between GH and PHASE to provide a discharge path for the HS MOSFET in the event that V_{CIN} goes to zero while V_{IN} is still applied.

Ground Connections (C_{GND} and P_{GND})

 $\mathsf{P}_{\mathsf{GND}}$ (power ground) should be externally connected to $\mathsf{C}_{\mathsf{GND}}$ (control signal ground). The layout of the printed circuit board should be such that the inductance separating $\mathsf{C}_{\mathsf{GND}}$ and $\mathsf{P}_{\mathsf{GND}}$ is minimized. Transient differences due to inductance effects between these two pins should not exceed 0.5 V

Control and Drive Supply Voltage Input (VDRV, VCIN)

 V_{CIN} is the bias supply for the gate drive control IC. V_{DRV} is the bias supply for the gate drivers. It is recommended to separate these pins through a resistor. This creates a low pass filtering effect to avoid coupling of high frequency gate drive noise into the IC.

Bootstrap Circuit (BOOT)

The internal bootstrap diode and an external bootstrap capacitor form a charge pump that supplies voltage to the BOOT pin. An integrated bootstrap diode is incorporated so that only an external capacitor is necessary to complete the bootstrap circuit. Connect a boot strap capacitor with one leg tied to BOOT pin and the other tied to PHASE pin.

Shoot-Through Protection and Adaptive Dead Time

The SiC788 and SiC788A have an internal adaptive logic to avoid shoot through and optimize dead time. The shoot through protection ensures that both high-side and low-side MOSFETs are not turned ON at the same time. The adaptive dead time control operates as follows. The high-side and low-side gate voltages are monitored to prevent the MOSFET turning ON from tuning ON until the other MOSFET's gate voltage is sufficiently low (< 1 V). Built in delays also ensure that one power MOSFET is completely OFF, before the other can be turned ON. This feature helps to adjust dead time as gate transitions change with respect to output current and temperature.

Under Voltage Lockout (UVLO)

During the start up cycle, the UVLO disables the gate drive, holding high-side and low-side MOSFET gates low, until the supply voltage rail has reached a point at which the logic circuitry can be safely activated. The SiC788 and SiC788A also incorporate logic to clamp the gate drive signals to zero when the UVLO falling edge triggers the shutdown of the device. As an added precaution, a 20 k Ω resistor is connected between GH and PHASE to provide a discharge path for the HS MOSFET.

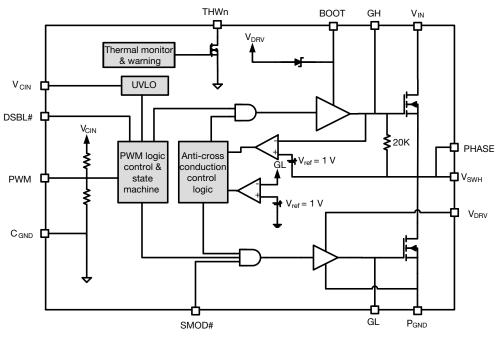


Fig. 3 - SiC788 and SiC788A Functional Block Diagram

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FUNCTIONAL BLOCK DIAGRAM



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DEVICE TRUTH TABLE							
DSBL#	SMOD#	PWM	GH	GL			
Open	Х	Х	L	L			
L	Х	Х	L	L			
Н	L	L	L	L			
Н	L	Н	Н	L			
Н	L	Tri-state	L	L			
Н	Н	L	L	Н			
Н	Н	Н	Н	L			
Н	Н	Tri-state	L	L			

PWM TIMING DIAGRAM

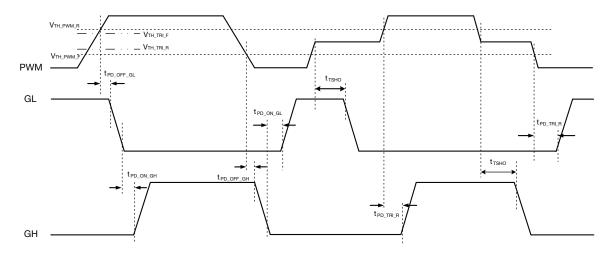
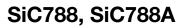
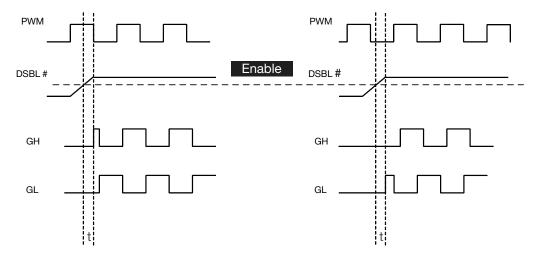


Fig. 4 - Definition of PWM Logic and Tri-State



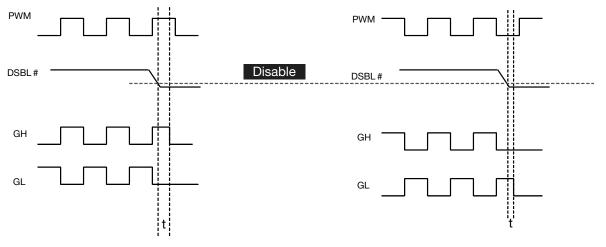


OPERATION TIMING DIAGRAM: DSBL#



DSBL# High to GH Rising Propagation Delay

DSBL# High to GL Rising Propagation Delay



DSBL# Low to GH Falling Propagation Delay

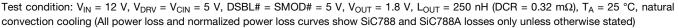
DSBL# Low to GL Falling Propagation Delay

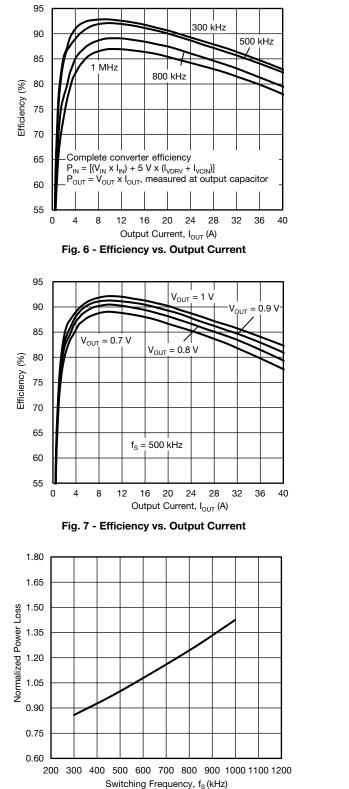
Fig. 5 - DSBL# Propagation Delay

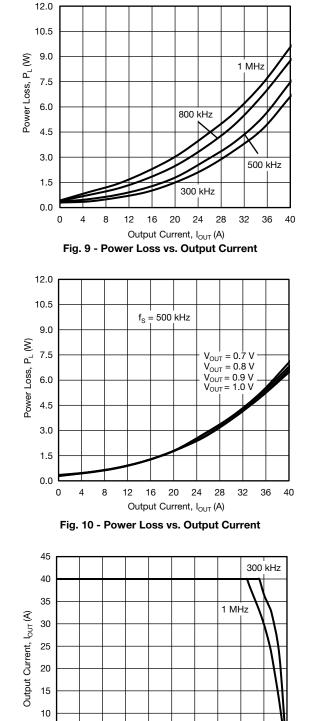
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ELECTRICAL CHARACTERISTICS







PCB Temperature, T_{PCB} (°C) Fig. 8 - Power Loss vs. Switching Frequency Fig. 11 - Safe Operating Area 9

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0 15 30 45 60 75 90

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105 120 135 150

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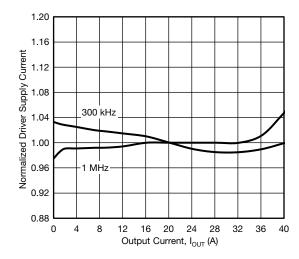


Fig. 12 - Driver Supply Current vs. Output Current

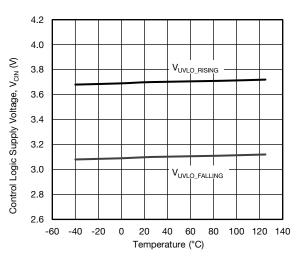


Fig. 13 - UVLO Threshold vs. Temperature

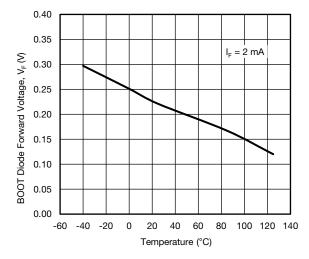


Fig. 15 - BOOT Diode Forward Voltage vs. Temperature

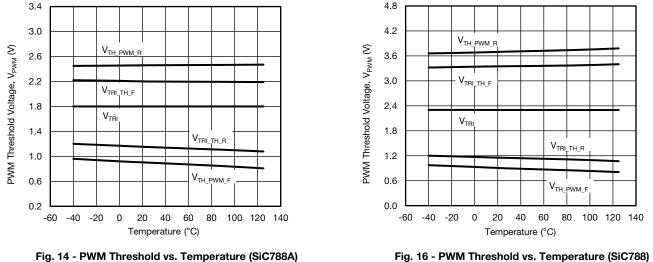


Fig. 16 - PWM Threshold vs. Temperature (SiC788)

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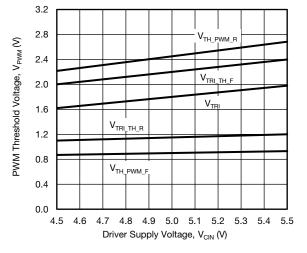


Fig. 17 - PWM Threshold vs. Driver Supply Voltage (SiC788A)

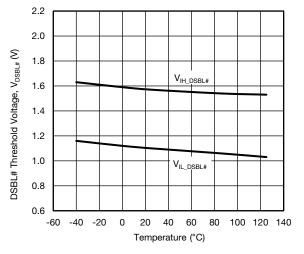


Fig. 18 - DSBL# Threshold vs. Temperature

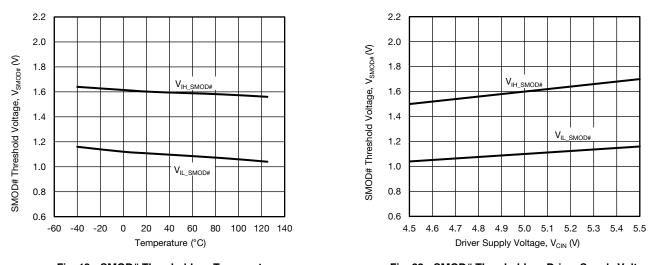


Fig. 19 - SMOD# Threshold vs. Temperature

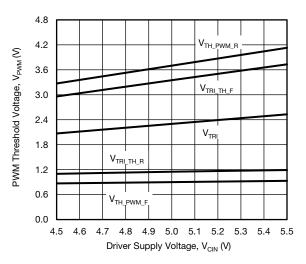


Fig. 20 - PWM Threshold vs. Driver Supply Voltage (SiC788)

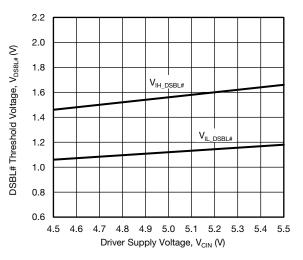
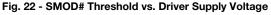


Fig. 21 - DSBL# Threshold vs. Driver Supply Voltage



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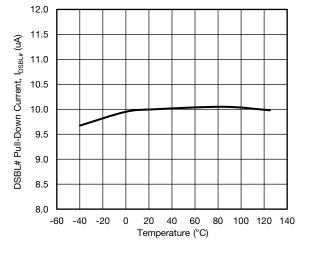


Fig. 23 - DSBL# Pull-down Current vs. Temperature

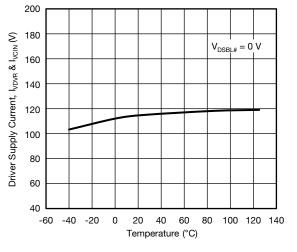


Fig. 24 - Driver Quiescent Current vs. Temperature

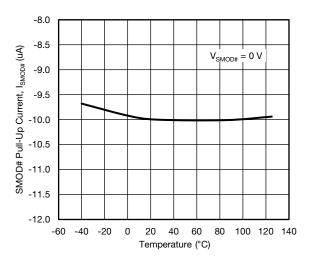


Fig. 25 - SMOD# Pull-up Current vs. Temperature

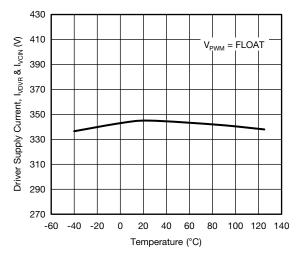
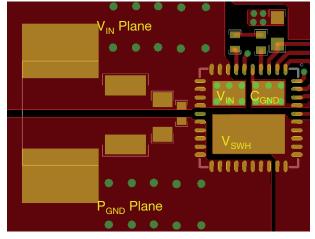


Fig. 26 - Driver Quiescent Current vs. Temperature



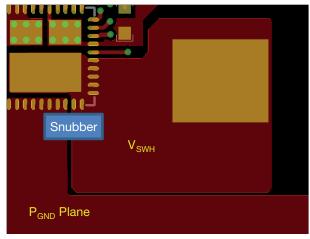
PCB LAYOUT RECOMMENDATIONS

Step 1: V_{IN} / P_{GND} Planes and Decoupling



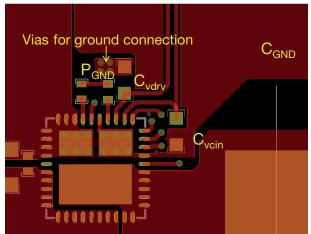
- 1. Layout V_{IN} and P_{GND} planes as shown above
- 2. Ceramic capacitors should be placed directly between V_{IN} and $P_{\text{GND}},$ and as close as possible to IC for best decoupling effect
- 3. Different ceramic capacitor values and packages should be used to cover entire decoupling spectrum, e.g. 1210, 0805, 0603, and 0402
- 4. Smaller capacitance values, placed closer to the IC's V_{IN} pin(s), result in better high frequency noise absorbing

Step 2: V_{SWH} Plane



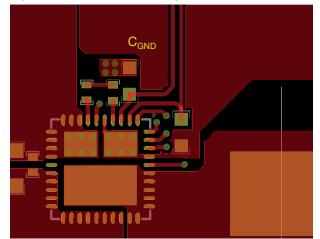
- 1. Connect output inductor to IC with large plane to lower resistance
- 2. V_{SWH} plane also serves as a heat-sink for low-side MOSFET. Please make the plane wide and short to achieve best thermal path
- 3. If a snubber network is required, place components as shown above

Step 3: V_{CIN} / V_{DRV} Input Filter



- 1. V_{CIN} / V_{DRV} input filter ceramic capacitors should be placed as close as possible to IC. It is recommended to connect two capacitors separately
- 2. V_{CIN} capacitor should be placed between pin 2 and pin 37 (C_{GND} of driver IC) to achieve best noise filtering
- 3. V_{DRV} capacitor should be placed between pin 3 and P_{GND} to provide maximum instantaneous driver current for low-side MOSFET during switching cycle. P_{GND} can be connected to inner ground plane through vias, as shown above
- 4. Pin 5 and pin 37 should be connected with $C_{\mbox{GND}}$ pad, as shown above
- 5. For connecting V_{CIN} to C_{GND} , it is recommended to use a large plane to reduce parasitic inductance

Step 4: BOOT Resistor and Capacitor Placement



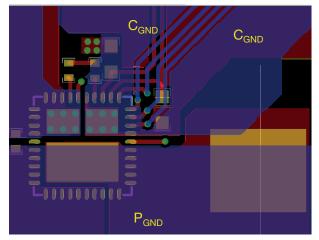
- 1. The components need to be placed as close as possible to IC, directly between PHASE (pin 7) and BOOT (pin 4)
- 2. To reduce parasitic inductance, 0402 package size can be used

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Step 5: Signal Routing



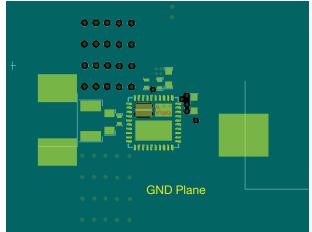
- 1. Route the PWM, SMOD#, DSBL#, and THWn signal traces out of the top right corner, next to pin 1
- 2. The PWM signal is a very important signal, both signal and return traces should not cross any power nodes on any layer
- 3. It is best to "shield" these traces from power switching nodes, e.g. $V_{\rm SWH},$ with a GND island to improve signal integrity

V_{IN} Plane

Step 6: Adding Thermal Relief Vias

- 1. Thermal relief vias can be added to the V_{IN} and C_{GND} pads to utilize inner layers for high-current and thermal dissipation
- 2. To achieve better thermal performance, additional vias can be added to V_{IN} and P_{GND} planes
- 3. The V_{SWH} pad is a noise source and it is not recommended to place vias on this pad
- 4. 8 mil vias for pads and 10 mils vias for planes are the optimal sizes. Vias on pad may drain solder during assembly and cause assembly issues. Please consult with the assembly house for guidelines

Step 7: Ground Connection



- 1. It is recommended to make the entire first inner layer (below top layer) the ground plane
- 2. The ground plane provides analog ground and power ground connections
- 3. The ground plane provides shielding between noise source on top layer and signal traces on bottom layer

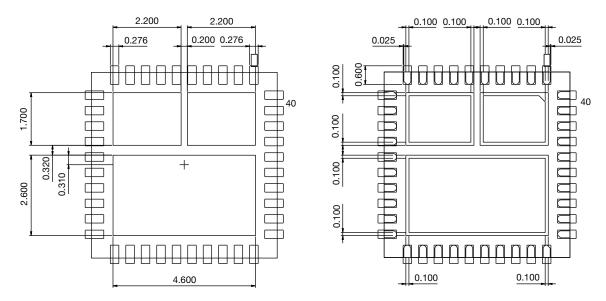


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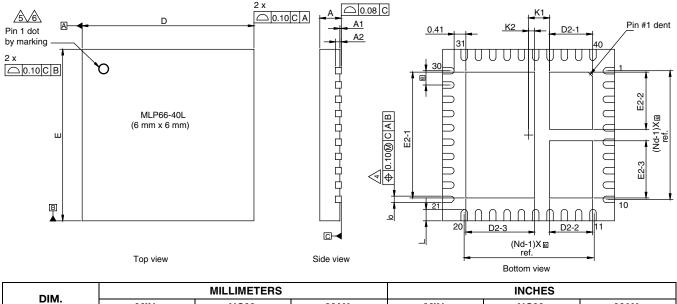
Vishay Siliconix

RECOMMENDED LAND PATTERN PowerPAK® MLP66-40L in millimeters





PACKAGE OUTLINE DRAWING MLP66-40L



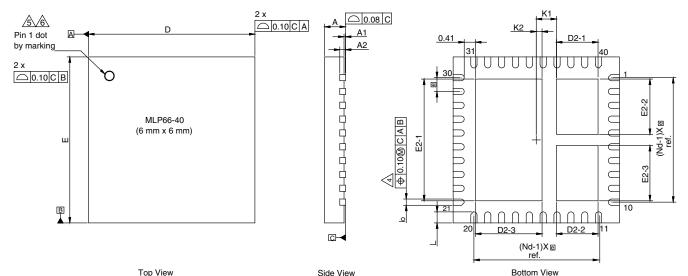
DIM.	MILLIMETERS			INCHES			
DIIVI.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	
А	0.70	0.75	0.80	0.027	0.029	0.031	
A1	0.00	-	0.05	0.000	-	0.002	
A2		0.20 ref.			0.008 ref.		
b	0.20	0.25	0.30	0.078	0.098	0.011	
D		6.00 BSC			0.236 BSC		
е		0.50 BSC			0.019 BSC		
E	6.00 BSC				0.236 BSC		
L	0.35	0.40	0.45	0.013	0.015	0.017	
Ν		40			40		
Nd		10		10			
Ne		10		10			
D2-1	1.45	1.50	1.55	0.057	0.059	0.061	
D2-2	1.45	1.50	1.55	0.057	0.059	0.061	
D2-3	2.35	2.40	2.45	0.095	0.094	0.096	
E2-1	4.35	4.40	4.45	0.171	0.173	0.175	
E2-2	1.95	2.00	2.05	0.076	0.078	0.080	
E2-3	1.95	2.00	2.05	0.076	0.078	0.080	
K1		0.73 BSC		0.028 BSC			
K2		0.21 BSC			0.008 BSC		

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Document Number: 62985



PowerPAK[®] MLP66-40 Case Outline



DIM.	MILLIMETERS			INCHES			
	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	
A ⁽⁸⁾	0.70	0.75	0.80	0.027	0.029	0.031	
A1	0.00	-	0.05	0.000	-	0.002	
A2		0.20 ref.			0.008 ref.		
b ⁽⁴⁾	0.20	0.25	0.30	0.078	0.098	0.011	
D		6.00 BSC			0.236 BSC		
е		0.50 BSC			0.019 BSC		
E		6.00 BSC			0.236 BSC		
L	0.35	0.40	0.45	0.013	0.015	0.017	
N ⁽³⁾		40		40			
Nd ⁽³⁾		10		10			
Ne ⁽³⁾		10		10			
D2-1	1.45	1.50	1.55	0.057	0.059	0.061	
D2-2	1.45	1.50	1.55	0.057	0.059	0.061	
D2-3	2.35	2.40	2.45	0.095	0.094	0.096	
E2-1	4.35	4.40	4.45	0.171	0.173	0.175	
E2-2	1.95	2.00	2.05	0.076	0.078	0.080	
E2-3	1.95	2.00	2.05	0.076	0.078	0.080	
K1		0.73 BSC			0.028 BSC		
K2	0.21 BSC			0.008 BSC			

Notes

1. Use millimeters as the primary measurement

2. Dimensioning and tolerances conform to ASME Y14.5M. - 1994

3. N is the number of terminals. Nd is the number of terminals in X-direction and Ne is the number of terminals in Y-direction

 Δ Dimension b applies to plated terminal and is measured between 0.20 mm and 0.25 mm from terminal tip

🛕 The pin #1 identifier must be existed on the top surface of the package by using indentation mark or other feature of package body

A Exact shape and size of this feature is optional

7. Package warpage max. 0.08 mm

Applied only for terminals

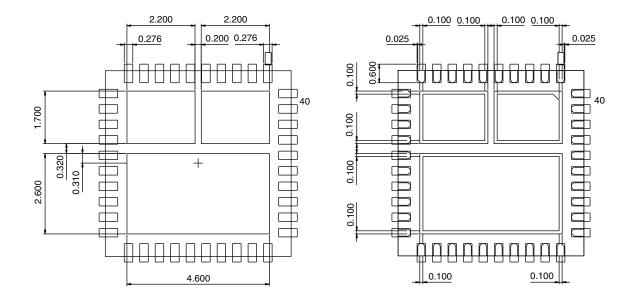
Revision: 12-Jan-15

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Recommended Land Pattern PowerPAK® MLP66-40L



All Dimensions are in milimeters



Vishay

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